

Outline

- Measurement Noise and Signal Conditioning ,
- Sources of Measurement Noise and techniques for its reduction;
- Analog Signal Filtering and processing operations.
- Grounding Techniques.

Noise and Interference

- Noise, by definition, is the presence of an unwanted electrical signal in a circuit.
- Interference is the undesirable effect of noise.
- Where a noise voltage causes improper operation of a circuit, or its relative magnitude is of the same order as the desired electrical signal, then it is interference.
- Noise itself cannot be totally eliminated but only reduced in magnitude until it no longer causes interference.
- This is especially true in data acquisition systems where the analog signal levels from transducers measuring a physical quantity can be very small.
- Compounding this in many instances is the physical cable distance over which these signals must be transmitted and the effect that noise may have on this extended circuitry.

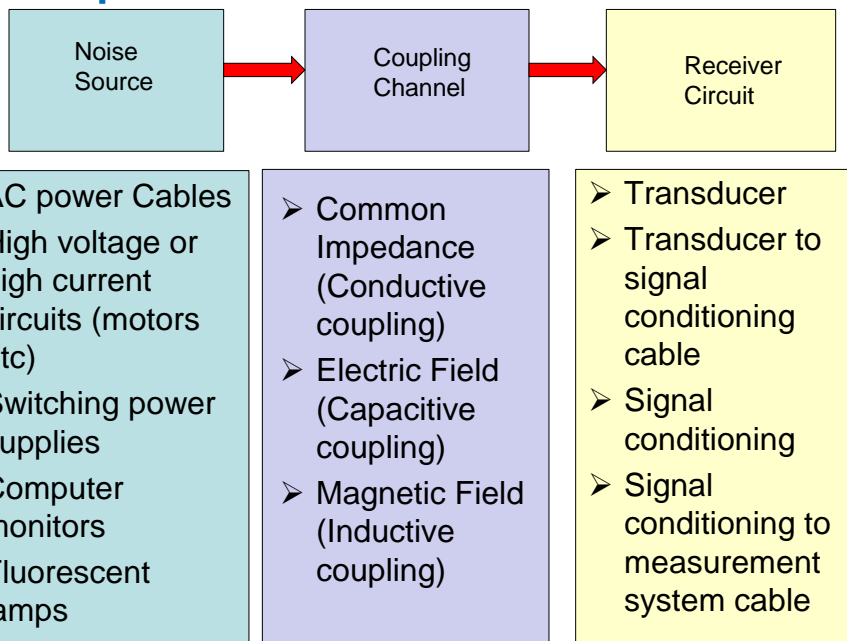
Noise and Interference

- Noise itself cannot be totally eliminated but only reduced in magnitude until it no longer causes interference.
- Noise is reduced by proper design practices and whatever noise that remain is treated by signal conditioning circuits
- Types of Noise :
 - Serial (differential type) which acts in series with useful output voltages of sensors and transducers and can cause significant errors
 - Common Mode noise which affects both lines in the same manner, it is dangerous since it can become differential in certain circumstances

Classification of Noise Sources

- **External Sources** such as motors, fluorescent lamps, monitors, mains cables, RF and audio-frequency circuits
- **Internal Sources** such as thermoelectric noise, shot noise and electrochemical action

Components of Noise induced Problem



Coupling Mechanisms

- The mechanisms for coupling noise most common to data acquisition and control applications are as follows:
 - Conductive coupling
 - Capacitive coupling
 - Inductive coupling
 - Other Coupling Mechanisms

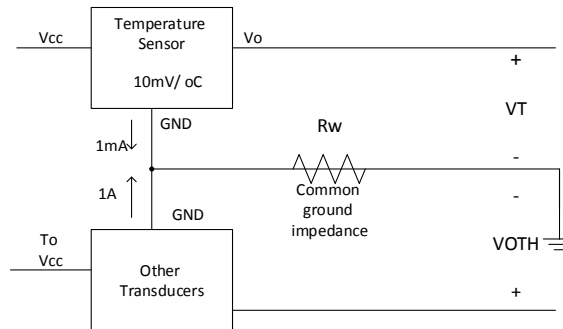
Conductive Coupling

- Conductive coupling occurs where two or more circuits share a common signal return.
- In such cases, return current from one circuit, flowing through the finite impedance of the common signal return, results in variations in the ground potential seen by the other circuits.

Conductive Coupling

- A series ground connection scheme resulting in conductive coupling is shown in Figure below.
- If the resistance of the common return lead is 0.1Ω and the return current from all other circuits is 1 A , then the voltage measured from the temperature sensor, (V_T), would vary by $0.1 \Omega \times 1 \text{ A} = 100 \text{ mV}$, corresponding to 10 degrees error in the temperature measured.

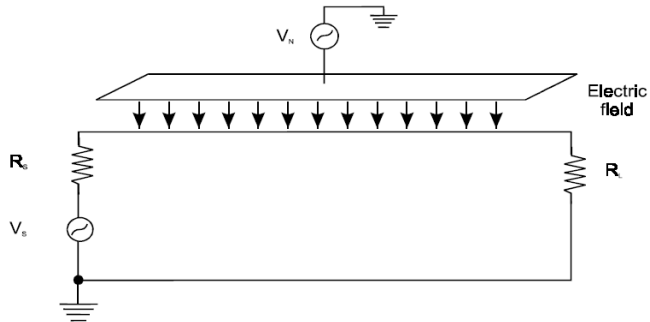
Series ground connections resulting in conductive coupling



Capacitive Coupling

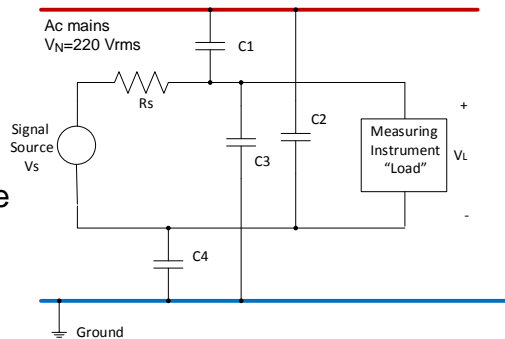
- Due to electric fields
- Electric fields occur in the vicinity of voltage-varying sources.
- Capacitive coupling **is the transmission of external noise through mutual and stray capacitances between a noise source and receiving circuit.**
- This is sometimes referred to as electrostatic coupling, although this is a misnomer, since the electrical fields are not static.
- Since cables tend to be the longest circuit elements, capacitive coupling is best demonstrated by considering a signal circuit connecting a signal source to a measurement system by a pair of long signal-carrying conductors.

- The physical representation of electric field coupling between a noise source and such a signal circuit is shown in Figure below.



Physical representation of an electrical field coupling into a signal circuit

- Noise voltage $V_n = \omega R_s C_n V_N$
- where C_n is any of the parasitic/ stray capacitors C_1, C_2, C_3, C_4
- If we make $C_1=C_2$ and $C_3=C_4$ then the noise voltage caused by these caps will be equal and opposite in sign and cancel each other
- Any unbalance in capacitor values will result in net noise voltage added or subtracted from V_L
- Also R_s must be smaller than load resistance and stray capacitance impedance



$$V_n \approx \omega R_s C_{12} V_N$$

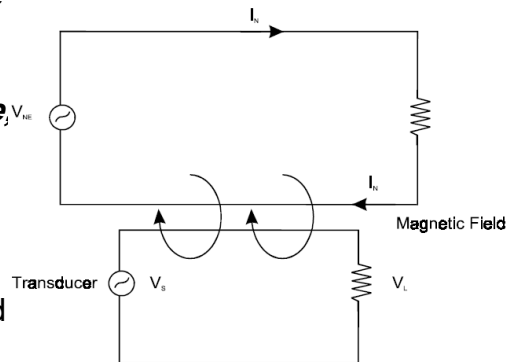
- ω , V_N - frequency and amplitude of the external noise source,
- R_s - the resistance to ground of the signal circuit
- C_{12} - and the mutual capacitance between them.
- If $R_s \gg 1/j\omega [C_{12} + C_{2G}]$, then it can be shown that the **capacitively-coupled noise voltage, is independent of the frequency of the noise source, and is much greater than in the case where the same resistance is relatively small.**

- The amplitude and the frequency of the noise source cannot be altered, the only means for reducing capacitive coupling into the signal circuit is to reduce the equivalent **signal circuit resistance to ground or reduce the mutual stray capacitance.**
- The mutual stray capacitance can be reduced by :
 - **increasing the relative distance of the signal wires from the noise source,**
 - **correct orientation of the conductors,**
 - **or by shielding.**

$$V_n \approx \omega R_s C_n V_N$$

Magnetic field coupling “Inductive Coupling”

- Magnetic field coupling or inductive coupling *is the mechanism by which time-varying magnetic fields produced by changing currents in a noise source link with current loops of receiving circuits.*
- The physical representation of magnetic field coupling between a noise source and a signal circuit is shown in Figure



- Lenz's law states that the voltage, V_n induced into a closed loop signal circuit of area A is proportional to the rate of change of the magnetic field coupling the circuit loop, the flux density (B) of the magnetic field and the area of the loop.
- This is represented by the formula:

$$V_n = 2 \pi f B A \cos \phi \quad (10^{-4})$$

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- where
 - f = the frequency of the sinusoidal varying flux density
 - B = the rms value of the flux density (gauss)
 - A = the area of the signal circuit loop (m^2)
 - ϕ = the angle between the flux density (B) and the area (A).
- This equation indicates that the noise voltage can be reduced by reducing B , A , or $\cos \phi$.
- The flux density (B) can be reduced by increasing the distance from the source of the field or if the field is caused by currents flowing through nearby pairs of wires, twisting those wires to reduce the net magnetic field effect to zero and or by alternating its direction.

Inductive Coupling

- The signal circuit loop area (A) can be reduced by placing the signal wires of the receiving circuit current loop closer together.
- For example, consider a signal circuit whose current carrying wires are 1 meter long and 1 centimeter apart, lying within a 10 gauss 60 Hz magnetic field, typical of fans, power wiring and transformers.
- The maximum voltage induced in the wires occurs for $\phi = 0^\circ$.
- $V_n = (2\pi \times 60)(10)(1 \times 10^{-2})(10^{-4}) = 3.7 \text{ mV}$.
- If the distance between the wires is reduced to 1 mm the noise voltage is reduced ten fold to 0.37 mV.

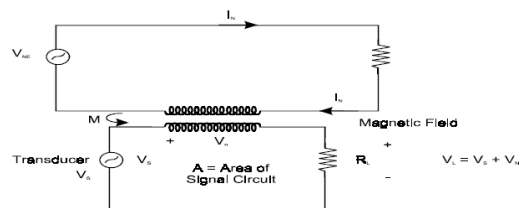
Inductive Coupling

- The $\cos\phi$, term can be reduced by correctly orienting the wires of the signal circuit in the magnetic field.
- For example, if the signal wires were perpendicular to the magnetic field ($\phi = 90^\circ$) the induced voltage could be reduced to zero, although practically this would not be possible.
- Running the signal wires together in the same cable as the wires carrying the noise current source would maximize the induced noise voltage

- The equivalent circuit model of magnetic coupling between a noise source and a signal circuit is shown in Figure below.
- In terms of the mutual inductance (M), V_n is given by:

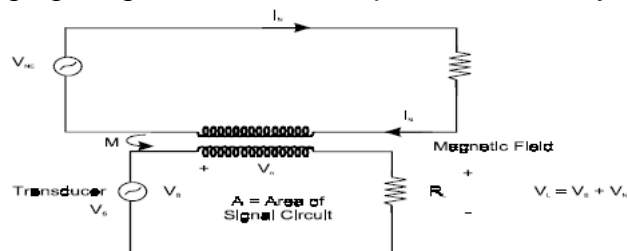
$$V_n = 2 \pi f M I_n$$

- I_n - is the rms value of the sinusoidal current in the noise circuit and f is its frequency.
- The mutual inductance (M) is directly proportional to the area (A) of the signal circuit current loop and the flux density, (B).



Inductive Coupling

- The physical geometry of the current loop of the receiving signal circuit, specifically its area, is the key to why it is susceptible to magnetic fields and how to minimize the effect.
- Cables provide the longest and largest current loop.
- The effect of magnetic coupling is best demonstrated by considering the circuit of Figure below, in which the signal cable current loop is coupled by a sinusoidal changing magnetic field with a peak flux density of $B\phi$.



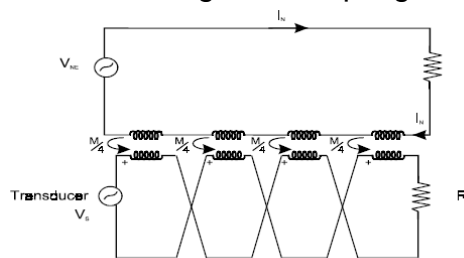
- Ideally, the only voltage appearing across the load should be V_s – the source signal voltage.
- However, the magnetic flux induces a voltage in the loop that appears in series with the receiver signal circuit.
- The voltage appearing across the load is the sum of the source voltage and the unwanted magnetic field induced voltage (V_N).

Techniques for Reduction of Noise

- 1. Location of signal wires:
 - Both mutual inductance and capacitance between signal wires are inversely proportional to the distance between them
 - It is recommended to place signal wires as far as possible from noise sources (at least 0.3 m)
- 2. Design of wires: Use twisted wires
- 3. Shielding
- 4. Proper grounding

Use twisted wires

- Twisting the insulated conductors together, can greatly reduce the amount of magnetic coupling into the signal lines.



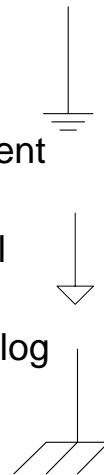
- The voltage induced in each section of the loop now alternates phases; its magnitude reduced by the reduction in area of each twisted loop (i.e. 1/4).
- Provided there is an even number of twists in the signal conductors, the voltages due to the magnetic field cancel out and only the desired signal voltage appears across the load.

Grounding Techniques

- The word ground has a historical origin that is, perhaps, the cause of the different meanings in use today.
- Originally, it referred to a point that was actually connected to earth in order to obtain zero potential
- In electronic systems, the ground point is the reference potential
- The confusion between earth and ground can be avoided if we consider that the electrical system on aircraft has a ground point for voltage reference, a point that is not connected to earth
- We will use earth for connection to the earth and ground as a central reference connection

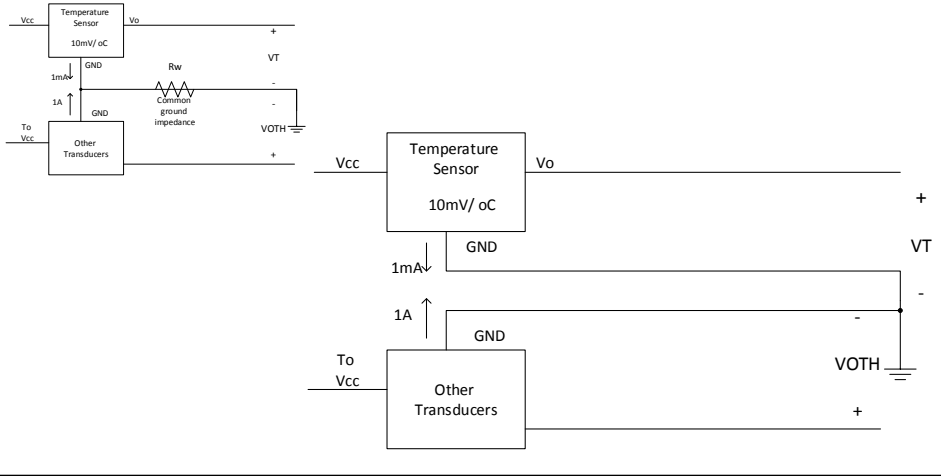
Grounding Techniques

- Usually we use the following Grounds:
- **Power Ground PGND** : provides path for current faults Logic
- **Digital Ground- DGND**: common for all digital and logic circuits
- **Analog Ground –AGND** : common for all analog signals
- **Safety Ground = Chassis Ground (Earth)** : connected to all metallic parts of the equipment to protect people if power lines come in contact with metal enclosure



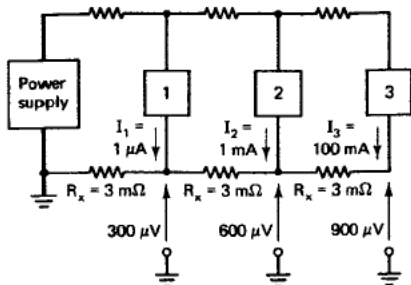
Solution to previous problem

- Don't allow the 1A current affect the output of the temperature sensor V_T



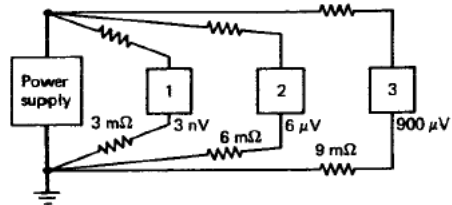
Ground Star Connection

Parallel distribution of power

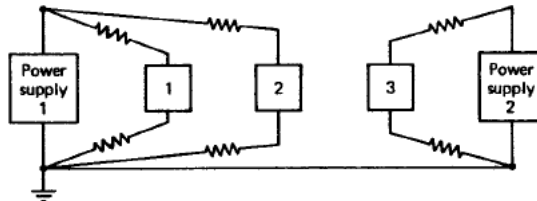


3 mohm is the resistance of 15 cm of AWG#18 wire

Radial (Star) distribution of power



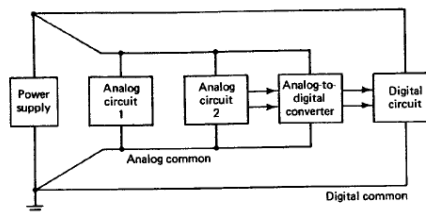
Two Power supplies can be used such that long wires are avoided for circuit 3



- If the voltage drop on power supply path does not affect the operation of the circuits, a combination of parallel and radial distribution could be used.
- The star connection will then be used for ground wire

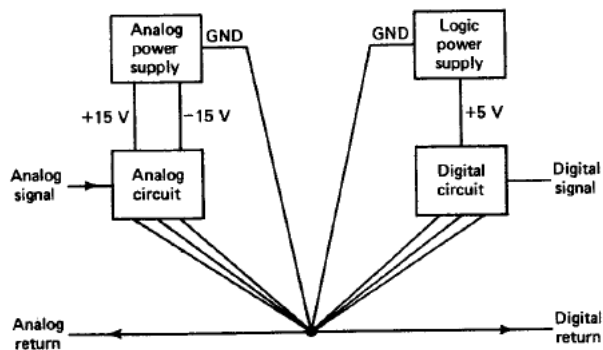
Grounding of Analog-Digital Circuits

- Consider the case when analog and digital circuits are used together
- Digital signals, create large current spikes in the ground paths due to switching
- These currents can cause much interference in analog circuits
- Even if they both share the same power supply, their ground wires must be different with only one common point as shown in the figure , this minimizes common impedances between digital and analog circuit



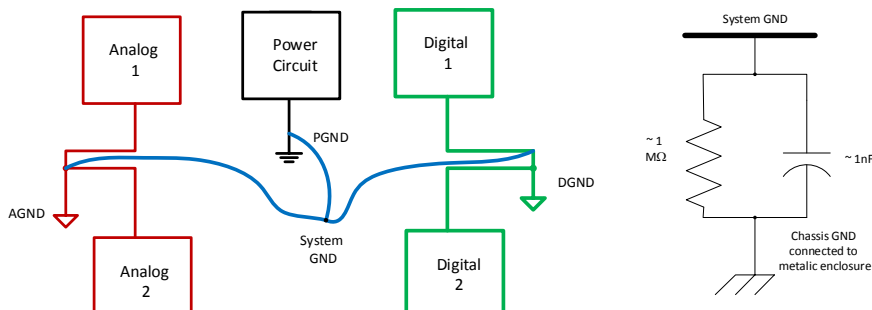
Analog-Digital Circuits

- When digital and analog circuits are powered by separate sources, each circuit must be connected to ground of its power supply.
- Then both grounds are connected to a single point ground that is called “star GND” or “system GND”



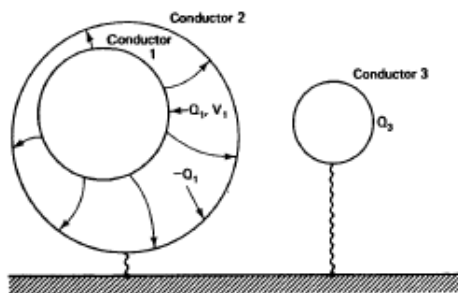
In schematic Capture (Board physical layout)

- Use Multiple grounds (with different symbols) according to type of circuit and then connect to the star point or system ground
- Connection of system ground to chassis (safety) ground can be done through a filter since system ground can act as a huge antenna that picks extra noise if connected to chassis directly, so a filter might be used



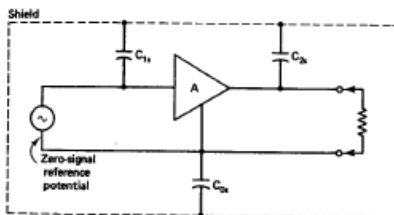
Electrical Shielding

- **Shielding:** adding a complete metallic enclosure or screen to remove capacitive coupling.
- Conductor 2 is the shield to prevent mutual capacitance between 1 and 3

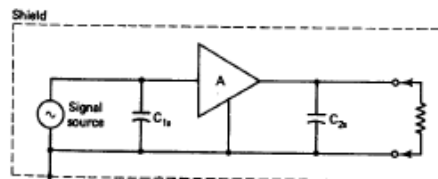


Rules of Shielding

- **Rule 1:** Connect the shield to low potential such as ground or earth, if not stray caps can appear as feedback caps affecting operation



Shield Floating
 C1s and C2s appear as Feedback Caps that can affect frequency response of amplifier



Shield Grounded
 C1s and C2s appear as input and output caps

- **Rule 2:** Shield conductor should be connected to zero-signal reference potential at the point the signal is earthed (point 1).
- Fig a shows an earthed signal source and an incorrect connection between the shield and the signal reference lead at some other point (3). There is no connection between the shield and zero reference at point 1
- Ground 1 and 2 usually at different potentials, or in other words, there is a ground voltage V_{12}

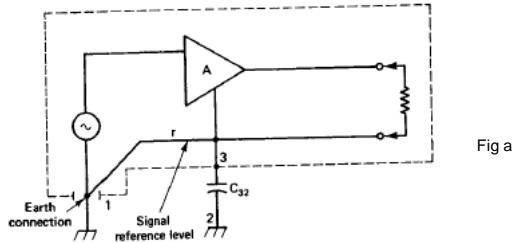


Fig a

- Fig b shows equivalent circuit of this case,
- The current due to V_{12} circulates through r , the reference signal wire, creating interference
- However if 1 and 3, besides being at the same potential, coincide in a single physical point, no current passes through r , and interference is eliminated (Fig c)

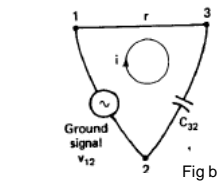


Fig b

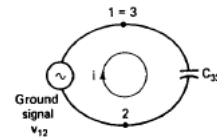


Fig c

Note that from electrical point of view Fig b and Fig c represent exactly the same circuit. However physical paths are different for the current

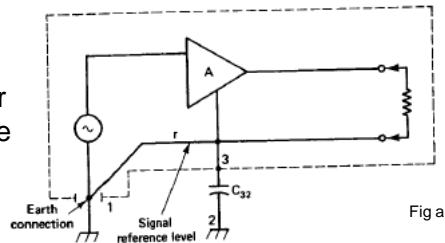


Fig a

Final Notes

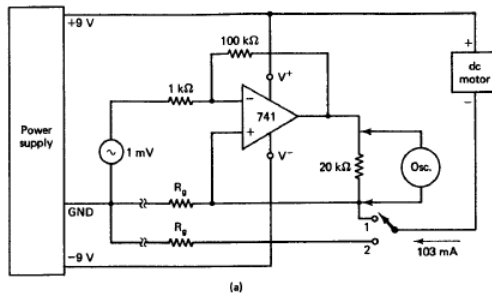
- Breaking the ground loop on the signal lines using transformers or optical couplers can provide additional noise reduction.
- The rules of shield grounding are as follows:
 - Where possible, cable shields should be earthed at one end only.
 - Where the source is ungrounded and the signal amplifier is grounded, the input shield should always be connected to the amplifier common terminal, even if this point is not at earth ground.
 - Where the source is grounded and the signal amplifier is ungrounded, the input shield should be connected to the source common terminal, even if this point is not at earth ground.
- Grounding the shield has additional benefits such as providing a path for RF currents and preventing the build-up of static charge by providing a discharge path to ground.

- Case Studies for reference only

Example 2.2

Figure 2.8(a) shows an experimental setup used to demonstrate the effectiveness of the star distribution of power. Two circuits, an amplifier with a gain of 100, and a dc tape recorder motor share the same power supply.

When the motor is connected to ground 1, its current (while returning to the power supply) causes a dc voltage drop across R_g of $240 \mu\text{V}$. A 24-mV dc level is measured across the load. The measured value for the motor return current is 103 mA, which indicates that the value for R_g is $2.3 \text{ m}\Omega$. By reducing the ground 1 wire to half its length, the dc level at the load becomes only 12 mV, indicating



that it is due to the voltage drop across the effective resistance of the ground wire, $R_g/2$ in this last case. When the motor is connected to ground 2, no dc level is measured at the load.

Grounds 1 and 2 have the same length (2 m) and they were built from the same wire (No. 22 stranded). The effects are due only to the different ground connection used. In the parallel distribution of power (ground 1) the return current

of the farther circuit (dc motor) affects the amplifier, which is closer to the source. The star distribution (ground 2) eliminates the effect of return currents through the same ground path.

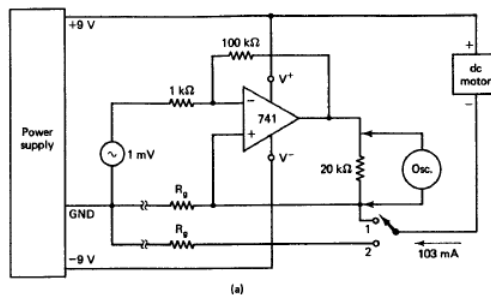
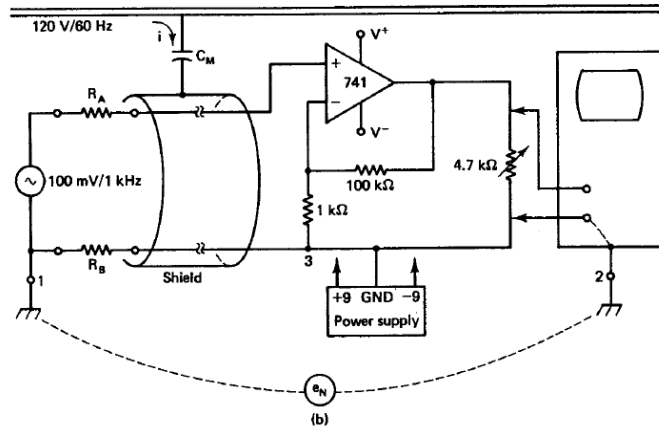
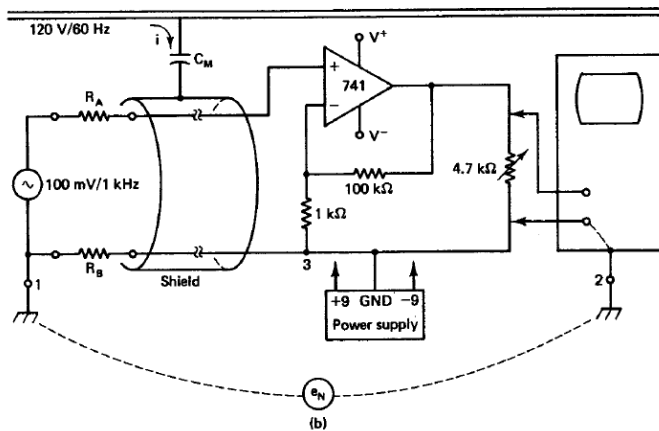


Figure 2.8(b) shows the circuit used to demonstrate the effectiveness and limitations of the previously described shielding techniques in the elimination of interference problems. A two-wire shielded cable connects a signal generator to a noninverting amplifier; this 2-m-long wire is placed near the power line (120 V/60 Hz) for higher electric field interference. Resistors R_A and R_B have been added to the signal wires in order to exaggerate interference effects. When electric field interference occurs, an interference current enters the signal line through the mutual capacitance C_M , causing on its way to ground a voltage drop across R_A .



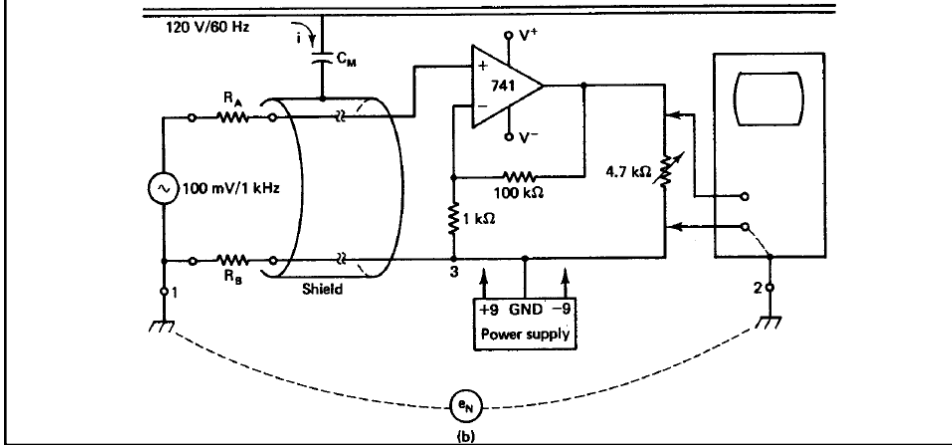
It is also possible to observe ground resistance interference when parts of the circuit are connected to different earth points. In this case, the potential difference between earths causes a current flow through the zero-signal-reference wire (ground loop), and the resulting voltage drop across R_B is amplified and observed in the oscilloscope. For a better display of the interference signals, the frequency of the source signal has been fixed at 1 kHz, quite distinguishable from 60-Hz noise.



The following cases show interference created by electric fields and different earth points. Interference is shown in the oscilloscope as a 60-Hz signal.

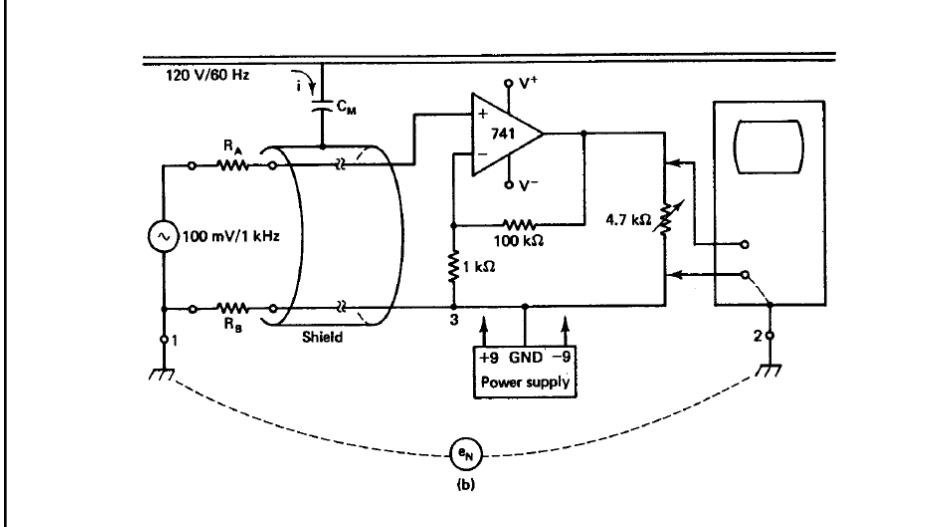
- a. $R_A = 0$, no shield, no noise
- b. $R_A = 100 \text{ k}\Omega$, no shield, 8-mV noise
- c. $R_A = 50 \text{ k}\Omega$, no shield, 4-mV noise

Cases b and c suggest that a 80-nA interference current flows through R_A to earth. These three cases apply when point 1 is earthed or unearthed. For the next cases



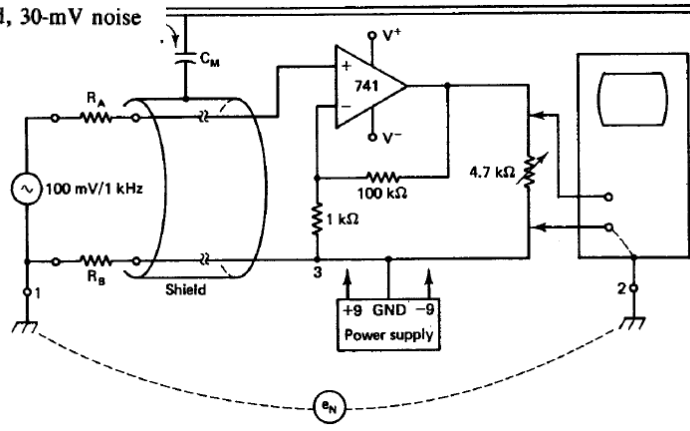
(d, e, and f), the shield is connected to point 1 or 3 (the zero-signal reference potential) and rule 1 of shielding applies.

- d. $R_A = 0$, no noise
- e. $R_A = 100 \text{ k}\Omega$, 2-mV noise, $i = 20 \text{ nA}$
- f. $R_A = 50 \text{ k}\Omega$, 1-mV noise, $i = 20 \text{ nA}$



If the signal (point 1) is earthed to the same earth point as the oscilloscope, $e_N = 0$ and interference is due only to electric fields. If points 1 and 2 are earthed to different earth points, $e_N \neq 0$ and the shield cannot eliminate the interference completely: two different types of interferences appear now, capacitive and resistive. The shield eliminates only the capacitive interference. The following cases apply for $e_N \neq 0$ (1 and 2 are different earth points).

- g. $R_B = 0$, no shield, 15-mV noise
- h. $R_B = 2 \text{ k}\Omega$, no shield, 40-mV noise
- i. $R_B = 0$, shielded, 10-mV noise
- j. $R_B = 2 \text{ k}\Omega$, shielded, 30-mV noise



The cases mentioned above are just a few among many possible cases of interference. The indicated noise levels were obtained experimentally and they may vary over a wide range; this is due mainly to the particular characteristics of the earth line during measurements (e.g., leakage currents, appliances connected to the line, etc.).

Nevertheless, and as a general conclusion of the experiment indicated above,

we may state that (1) shielding eliminates only interference due to electric field effects, (2) having only one earthed point eliminates ground loops, and (3) interference increases for higher effective values of the resistance along signal lines.